

# Improvement of SGP modelling using additional devices

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#### **Motivation**

- NXP small signal bipolar transistors are modeled by using Standard Gummel Poon (SGP) model
- Some models offered to customers are old and no more accurate (e.g. by redesign of die)
- > Revision of these models
- > SGP very simple model compared with HICUM etc. but:



## **Motivation**

- Many customers like SGP model because of its properties:
  - simple model
  - cheap and simple simulators (e.g. PSpice etc.) available
- > For many small signal transistors SGP model is sufficient
- > Some disadvantages, especially poor fitting in some kinds of operation
- Misfitting often occurs when modelling transistors with higher VCEO (typically with VCEO > 40V)



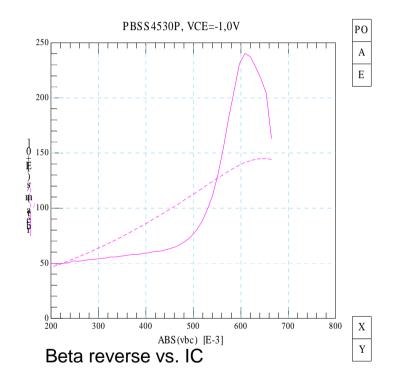
# Modelling problem (1)

Modelling the BF Slope without additional devices is difficult

Poor fitting of reverse BF is hampering forward modelling

Solid line: measured data

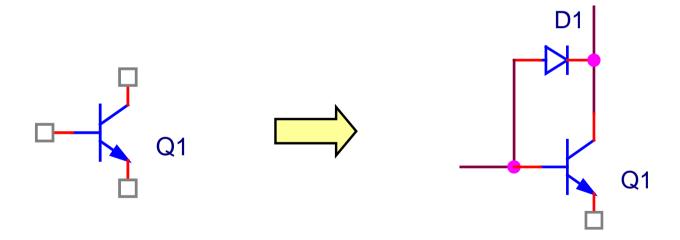
Dashed line: simulated data





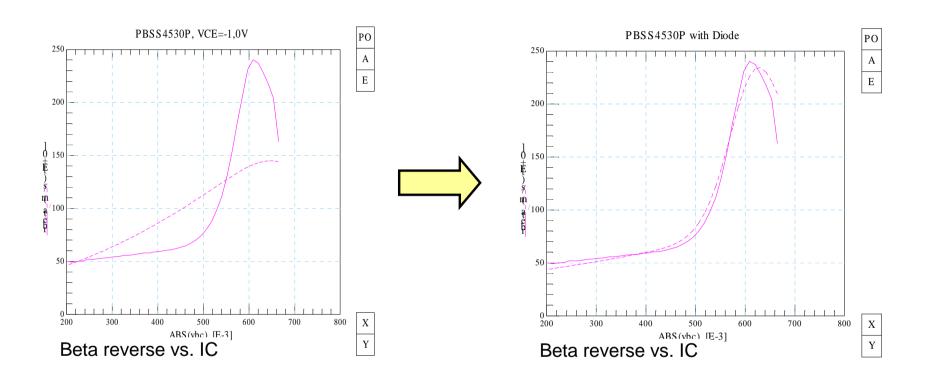
## Improvement of reverse modelling

Use of a additional diode parallel to the B-C junction of the transistor in conducting direction





# Improvement of reverse modelling



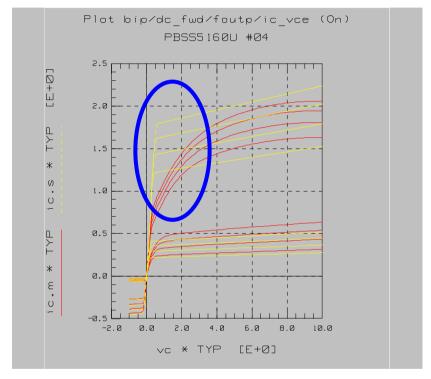
- > Result: Good fitting in reverse mode of operation
- > Facilitation of modeling in forward direction



# Modelling problem (2)

- SGP-Model is not able to calculate area of quasisaturation accurately
- No smooth change, only sharp bend
- Poor fitting in this area

Red line: measured data Yellow line: simulated data



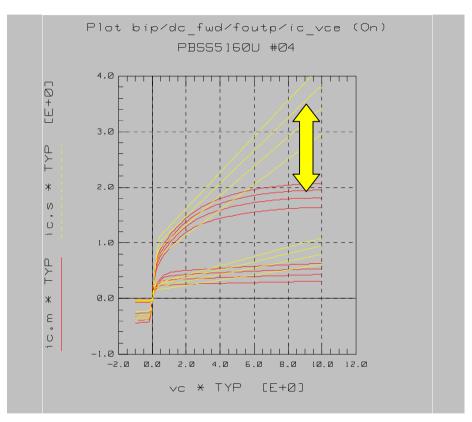
output characteristic (VC vs. IC)



## Modelling up to now

- Decreasing of VAF (and IKF) to archive better fitting of quasi saturation
- Possible solution for low-VCEsattransistors, but poor fitting in area of linear operation!

Red line: measured data Yellow line: simulated data

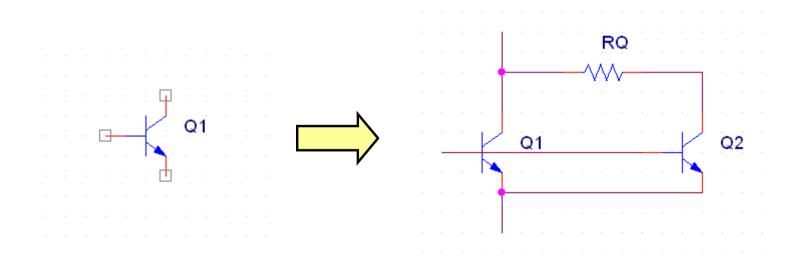


output characteristic (VC vs. IC)



# Improvement of quasi saturation

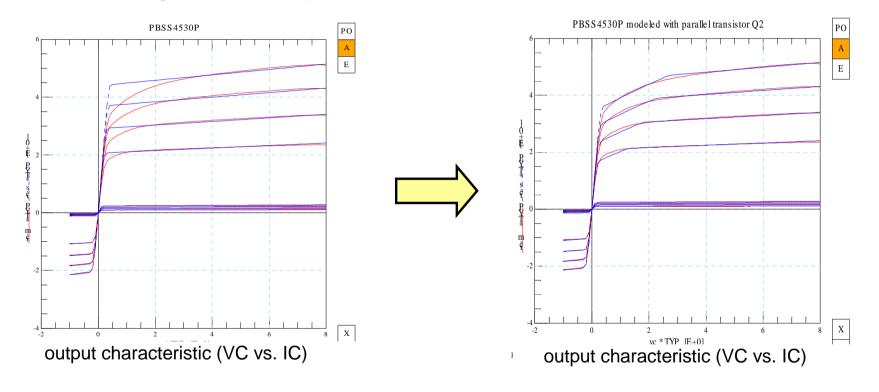
 Use of a second transistor (Q2) with resistor (RQ) parallel to the main transistor (Q1)





## Improvement of quasi saturation

➤ Modelling result with parallel transistor Q2:

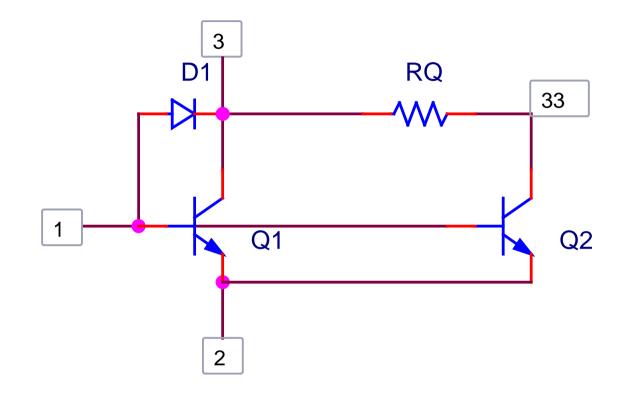


Red line: measured data Blue line: simulated data



## **Example for enhanced model**

.SUBCKT PBSS4612PA 3 1 2 Q1 3 1 2 MAIN + Ares - 0.5663 Q2 33 1 2 MAIN + Arex - 0.03168 D1 1 3 DIODE RCQ 3 33 11.23 .MODEL MAIN NPN + IS = 3.467E-012+ NF = 0.9647+ etc. .MODEL DIODE D + IS = 1.313E-014 + N = 1.01+ etc... .ENDS



Ratio of Q1 and Q2 is determined by the variable Area



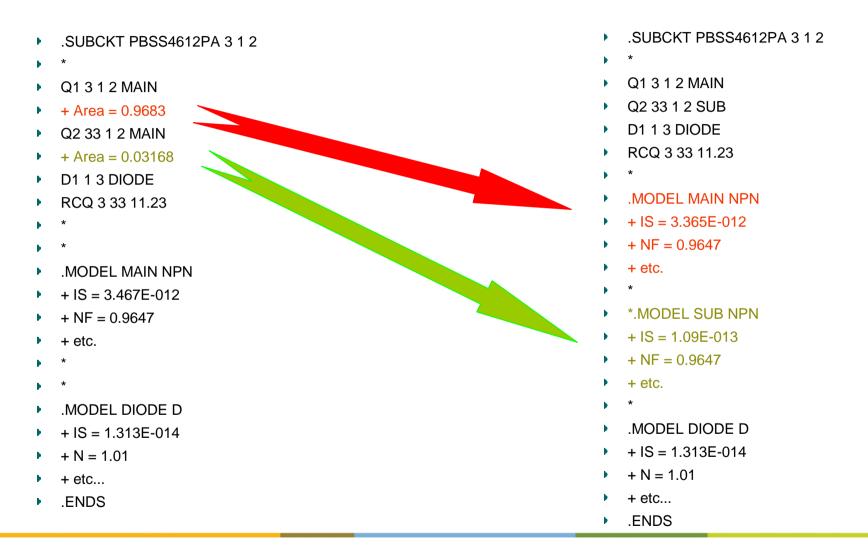
## **Difficulties**

- > Some simulators (e.g. PSpice student) do no accept the variable "Area"
- > Possible solution: Use of two different transistors with different parameters and "included" area-variable:
- > Proceeding:
  - To be multiplied with area: IS, ISE, ISC, ISS, IKF, IKR, IRB, ITF, CJE, CJC, CJS, QCO
  - To be divided by area:RE, RC, RB, RBM
  - All other Parameters stay unchanged

(Source: Khakzar, ISBN 3-8169-1262-1, page 222)



#### Parameters for new model



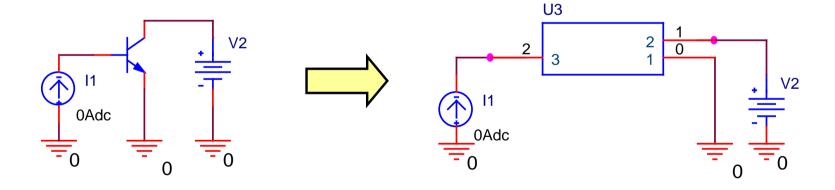


## Remarks

Disadvantage:

In schematics subcircuits are displayed as black box; the transistor symbol will be lost.

> Possibility of creating an .olb-file, but circumstancial





## **Summary**

- Modelling problems with bipolar transistors with higher VCEO using SGP-model
- Quasi saturation and reverse mode of operation can be improved by inserting additional diode and transistor / resistor
- > Necessity of subcircuits
- > Higher accuracy of models, but even higher complexity



#### Remarks

- Experience with such complex models? Most companies are offering simple models without using subcircuits. => reason?
- > Difficult to get customer feedback about such matters.
- > Acceptance of customers especially concerning confusing schematics?



